



**SGMICRO**

# SGM8607-2 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## GENERAL DESCRIPTION

The SGM8607-2 integrates high precision, low noise, low voltage and low power dual CMOS operational amplifiers in a small, 9-ball WLCSP package. It can operate from 1.8V to 5.5V power supply over the -40°C to +125°C temperature range or operate from 1.7V to 5.5V power supply over the 0°C to +70°C temperature range. The SGM8607-2 supports rail-to-rail input and output operation. The input common mode voltage range is from  $(-V_S) - 0.1V$  to  $(+V_S) + 0.1V$ , and the output range is from  $(-V_S) + 0.032V$  to  $(+V_S) - 0.05V$  with 600Ω load resistor.

The SGM8607-2 features an input offset voltage of 30μV (MAX), a gain-bandwidth product of 1.2MHz, and a slew rate of 0.6V/μs.

The SGM8607-2 is available in a Green WLCSP-1.2×1.2-9B-A package. It is specified over the extended industrial temperature range (-40°C to +125°C).

## APPLICATIONS

Sensors  
Audio  
Active Filters  
A/D Converters  
Communications  
Test Equipment  
Cellular and Cordless Phones  
Laptops and PDAs  
Photodiode Amplification

## FEATURES

- **Low Input Offset Voltage: 30μV (MAX)**
- **Low Input Bias Current: 70pA (TYP)**
- **Input Voltage Noise Density: 27nV/√Hz at 10kHz**
- **Gain-Bandwidth Product: 1.2MHz**
- **Gain = +20 Stable for  $C_L = 1nF$   
Unity-Gain Stable for  $C_L = 100pF$**
- **Slew Rate: 0.6V/μs**
- **Settling Time to 0.1% with 2V Step: 1μs**
- **Overload Recovery Time: 0.5μs**
- **Rail-to-Rail Input and Output**
- **Power Supply Voltage:**
  - ♦ **Low 1.8V Supply Rail over the -40°C to +125°C Range**
  - ♦ **Low 1.7V Supply Rail over the 0°C to +70°C Range**
  - ♦ **High Supply Voltage: 5.5V**
- **Input Signal Range:  $(-V_S) - 0.1V$  to  $(+V_S) + 0.1V$**
- **Low Power:**
  - ♦ **140μA (TYP)**
  - ♦ **0.6μA (MAX) Supply Current in Shutdown**
- **Available in a Green WLCSP-1.2×1.2-9B-A Package**

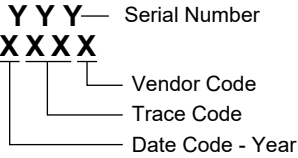
# 1.2MHz, High Precision, Rail-to-Rail I/O, Dual SGM8607-2 CMOS Operational Amplifier in WLCSP Package

## PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION
SGM8607-2	WLCSP-1.2x1.2-9B-A	-40°C to +125°C	SGM8607-2XG/TR	011 XXXX	Tape and Reel, 3000

## MARKING INFORMATION

NOTE: XXXX = Date Code, Trace Code and Vendor Code.



Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage, +V <sub>s</sub> to -V <sub>s</sub> .....	6.0V
Input Common Mode Voltage Range .....	(-V <sub>s</sub> ) - 0.3V to (+V <sub>s</sub> ) + 0.3V
Junction Temperature .....	+150°C
Storage Temperature Range .....	-65°C to +150°C
Lead Temperature (Soldering, 10s) .....	+260°C
ESD Susceptibility	
HBM .....	4000V
CDM .....	1000V

## RECOMMENDED OPERATING CONDITIONS

Operating Temperature Range .....	-40°C to +125°C
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## OVERSTRESS CAUTION

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

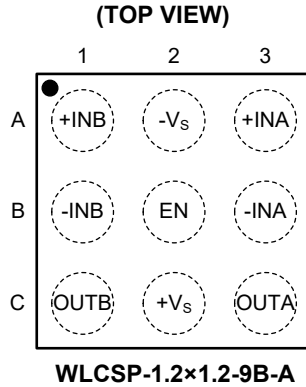
## ESD SENSITIVITY CAUTION

This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

## DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

**PIN CONFIGURATION**



**PIN DESCRIPTION**

PIN	NAME	FUNCTION
A1	+INB	Non-Inverting Input of Amplifier B.
A2	-Vs	Negative Power Supply.
A3	+INA	Non-Inverting Input of Amplifier A.
B1	-INB	Inverting Input of Amplifier B.
B2	EN	Active High Enable Input. When EN = High, the SGM8607-2 is in active status. When EN = Low, the SGM8607-2 is in shutdown status.
B3	-INA	Inverting Input of Amplifier A.
C1	OUTB	Output of Amplifier B.
C2	+Vs	Positive Power Supply.
C3	OUTA	Output of Amplifier A.

# SGM8607-2

## 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

### ELECTRICAL CHARACTERISTICS

( $V_S = 3.3V$ ,  $V_{CM} = V_S/2$ ,  $V_{EN} = +V_S$ , Full =  $-40^\circ C$  to  $+125^\circ C$ , typical values are at  $T_A = +25^\circ C$ , unless otherwise noted.)

PARAMETER	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS	
<b>Input Characteristics</b>							
Input Offset Voltage ( $V_{OS}$ )	$V_S = 5V$	+25°C		5	30	μV	
		Full			90		
Input Offset Voltage Drift ( $\Delta V_{OS}/\Delta T$ )		Full		0.12	0.6	μV/°C	
Input Bias Current ( $I_B$ )	$V_S = 1.8V$ to $5.5V$	+25°C		70	2000	pA	
		Full			9	nA	
Input Offset Current ( $I_{OS}$ )	$V_S = 1.8V$ to $5.5V$	+25°C		140		pA	
Input Common Mode Voltage Range ( $V_{CM}$ )		Full	$(-V_S) - 0.1$		$(+V_S) + 0.1$	V	
Common Mode Rejection Ratio (CMRR)	$(-V_S) - 0.1V \leq V_{CM} \leq (+V_S) + 0.1V$	+25°C	98	114		dB	
		Full	95				
Open-Loop Voltage Gain ( $A_{OL}$ )	$(-V_S) + 0.25V < V_{OUT} < (+V_S) - 0.25V$ , $R_L = 600\Omega$	+25°C	101	114		dB	
		Full	98				
	$(-V_S) + 0.1V < V_{OUT} < (+V_S) - 0.1V$ , $R_L = 10k\Omega$	+25°C	102	120			
		Full	99				
Input Capacitance ( $C_{IN}$ )		+25°C		30		pF	
<b>Output Characteristics</b>							
Output Voltage Low	$V_{OL} = V_{OUT} - (-V_S)$	$R_L = 600\Omega$	+25°C		32	50	mV
			Full			70	
		$R_L = 10k\Omega$	+25°C		2	5	
			Full			7	
Output Voltage High	$V_{OH} = (+V_S) - V_{OUT}$	$R_L = 600\Omega$	+25°C		50	65	mV
			Full			92	
		$R_L = 10k\Omega$	+25°C		3	7	
			Full			9	
Output Short-Current Limited ( $I_{LIM}$ )	Short to $+V_S$ or $-V_S$	+25°C		28		mA	
<b>Power Supply</b>							
Operating Voltage Range ( $V_S$ )		Full	1.8		5.5	V	
		+25°C	1.7		5.5		
Power Supply Rejection Ratio (PSRR)	$V_S = 1.8V$ to $5.5V$ , $V_{CM} = (-V_S) + 0.5V$	+25°C	100	124		dB	
		Full	97				
Quiescent Current ( $I_Q$ )	$I_{OUT} = 0A$	+25°C		140	195	μA	
		Full			235		
Shutdown Supply Current ( $I_{SHDN}$ )	$I_{OUT} = 0A$ , $V_{EN} = -V_S$	+25°C			0.6	μA	

# 1.2MHz, High Precision, Rail-to-Rail I/O, Dual SGM8607-2 CMOS Operational Amplifier in WLCSP Package

## ELECTRICAL CHARACTERISTICS (continued)

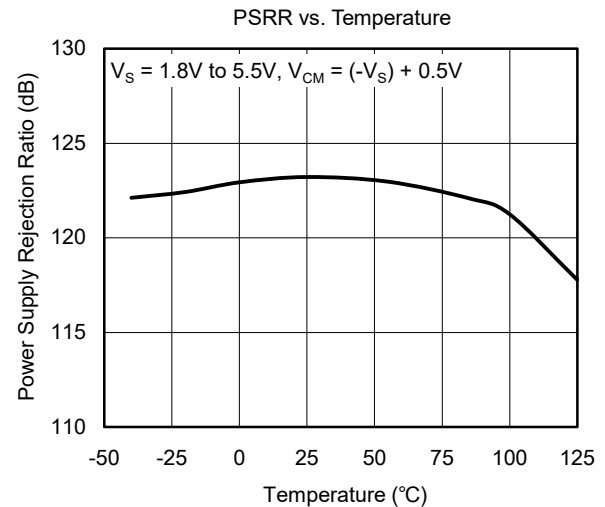
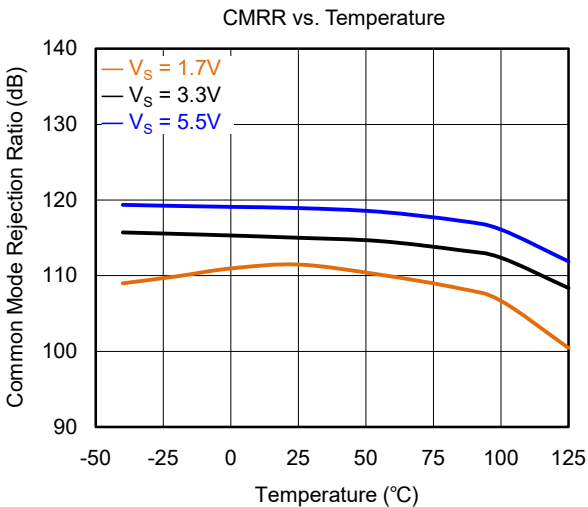
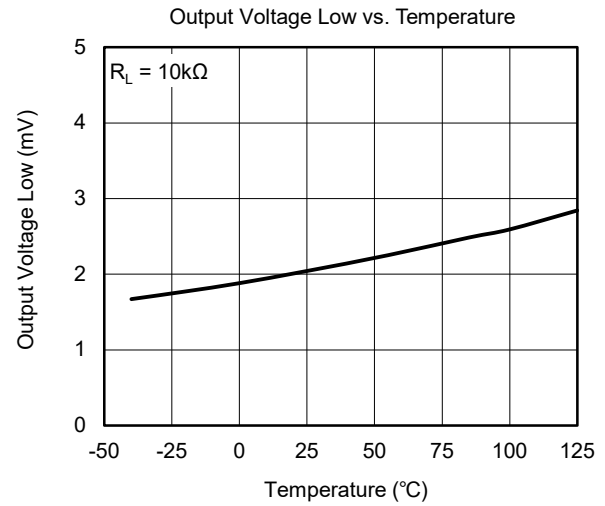
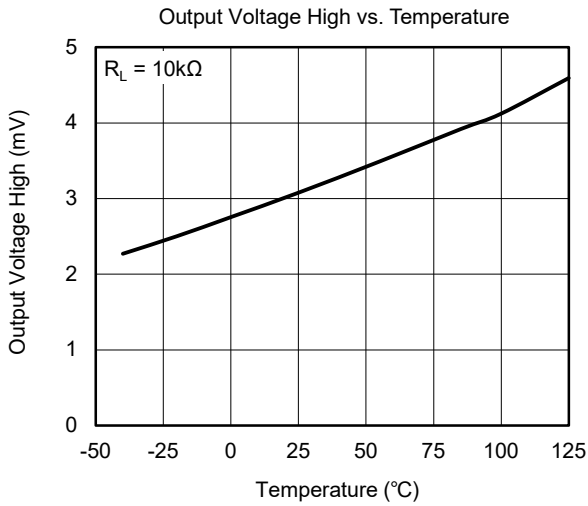
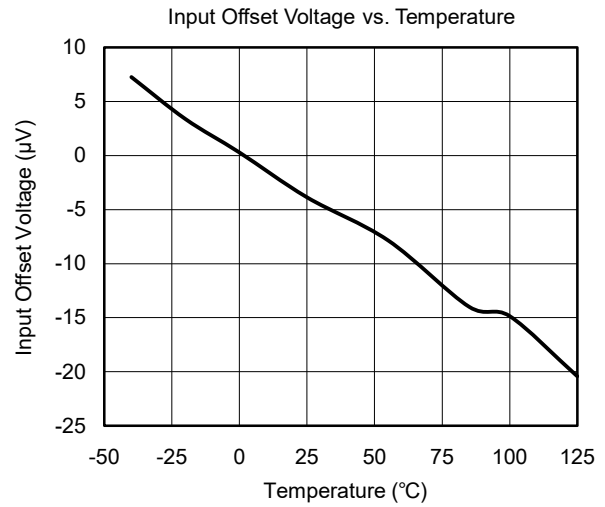
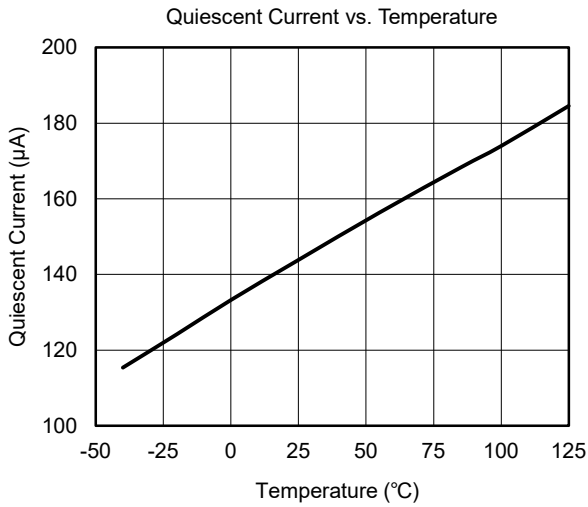
( $V_S = 3.3V$ ,  $V_{CM} = V_S/2$ ,  $V_{EN} = +V_S$ , Full =  $-40^\circ C$  to  $+125^\circ C$ , typical values are at  $T_A = +25^\circ C$ , unless otherwise noted.)

PARAMETER	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS
<b>Dynamic Performance</b>						
Gain-Bandwidth Product (GBP)		+25°C		1.2		MHz
Phase Margin ( $\phi_o$ )	G = +20, $R_L = 10k\Omega$ , $C_L = 1nF$	+25°C		56		°
	G = +1, $R_L = 10k\Omega$ , $C_L = 100pF$	+25°C		55		
Slew Rate (SR)	G = +1, 2V output step	+25°C		0.6		V/ $\mu s$
Settling Time to 0.1% ( $t_s$ )	G = +1, 2V output step	+25°C		1		$\mu s$
Overload Recovery Time (ORT)	$V_{IN} \times G = V_S$	+25°C		0.5		$\mu s$
<b>Noise Performance</b>						
Input Voltage Noise	f = 0.1Hz to 10Hz	+25°C		0.4		$\mu V_{P-P}$
Input Voltage Noise Density ( $e_n$ )	f = 10kHz	Full		27	55	nV/ $\sqrt{Hz}$
<b>EN Control</b>						
Input Voltage Low ( $V_{IL}$ )	+ $V_S = 1.8V$ to 5.5V, - $V_S = GND$	Full			0.4	V
Input Voltage High ( $V_{IH}$ )	+ $V_S = 1.8V$ to 5.5V, - $V_S = GND$	Full	1.7			V
Turn-On Time from Shutdown ( $t_{SHDN}$ )		+25°C		275	420	$\mu s$
		Full			600	
Power-Up Time ( $t_{ON}$ )		Full		0.6	1.2	ms

# SGM8607-2 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## TYPICAL PERFORMANCE CHARACTERISTICS

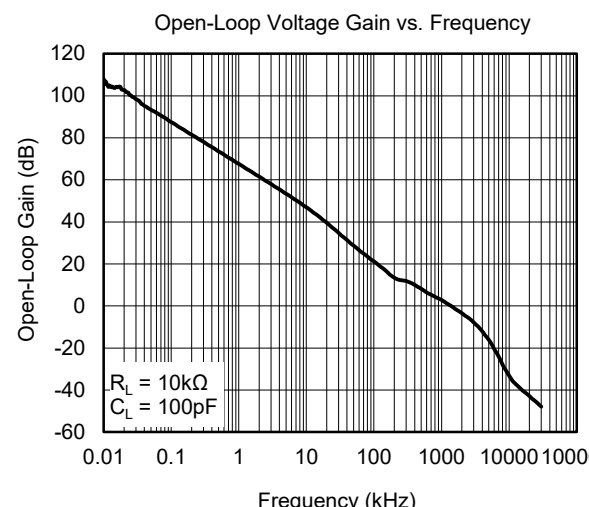
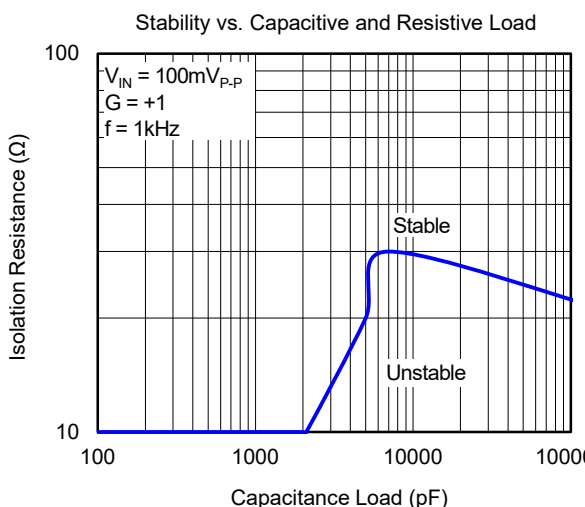
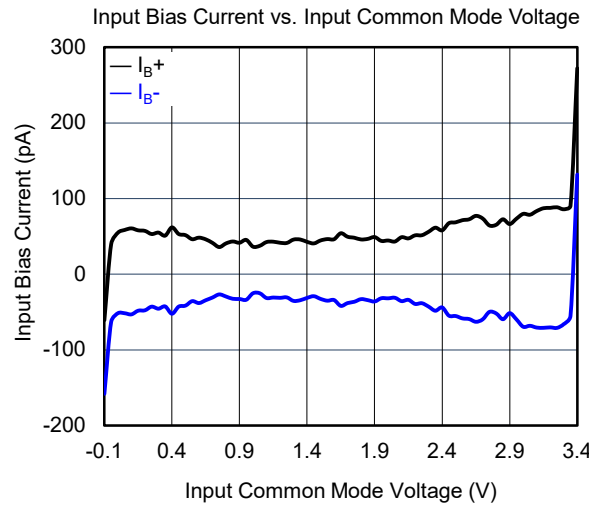
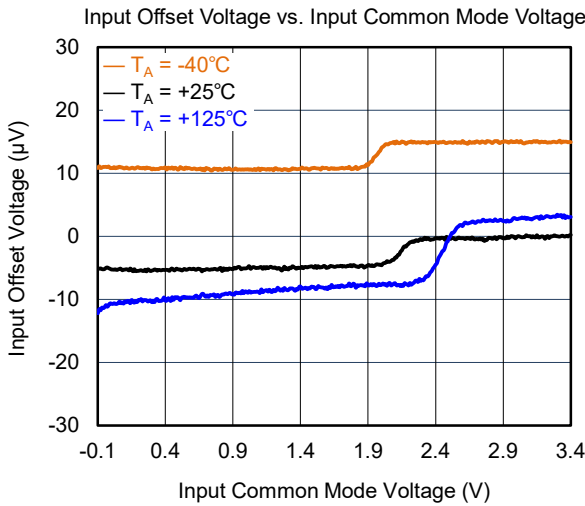
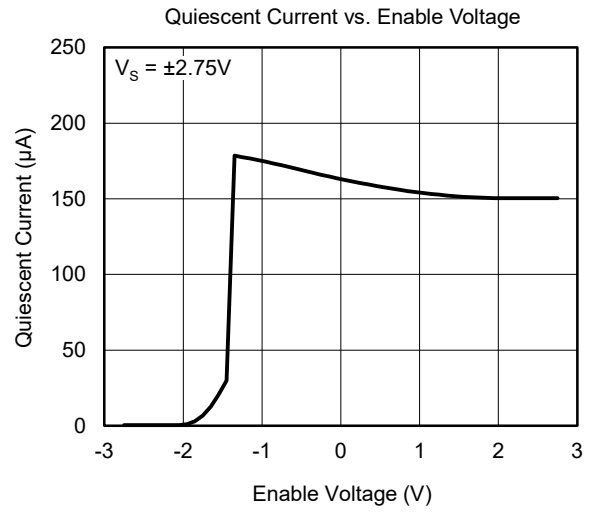
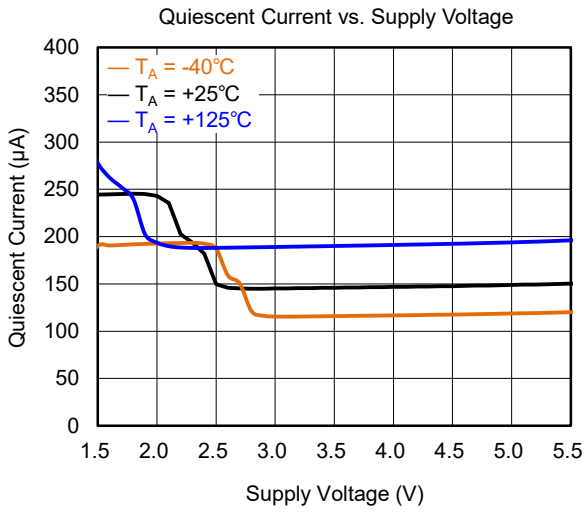
At  $T_A = +25^\circ\text{C}$ ,  $V_S = 3.3\text{V}$ ,  $V_{CM} = V_S/2$ ,  $V_{EN} = +V_S$ , unless otherwise noted.



# SGM8607-2 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

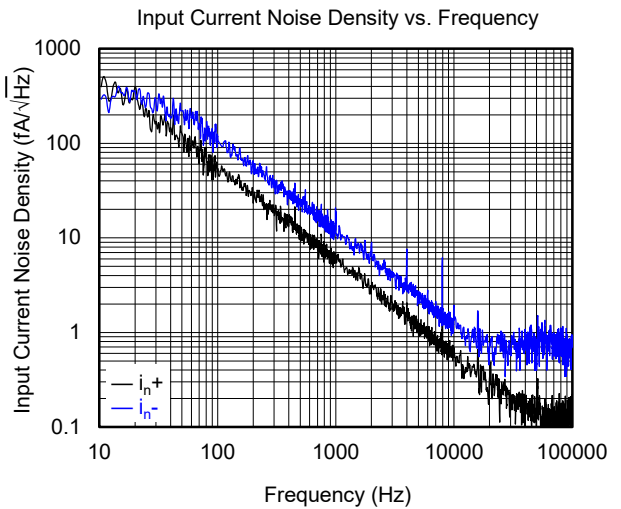
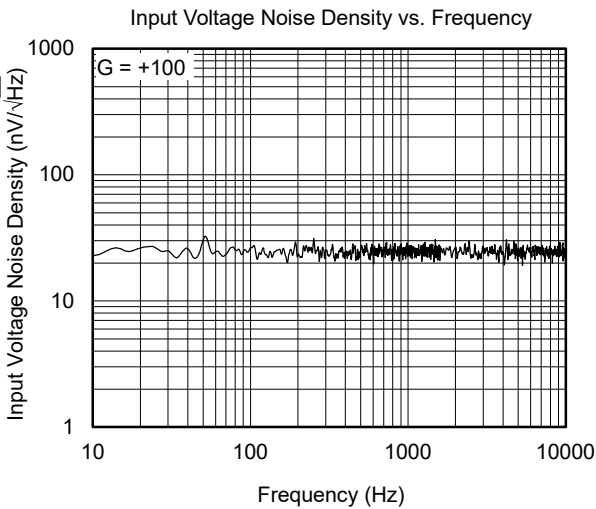
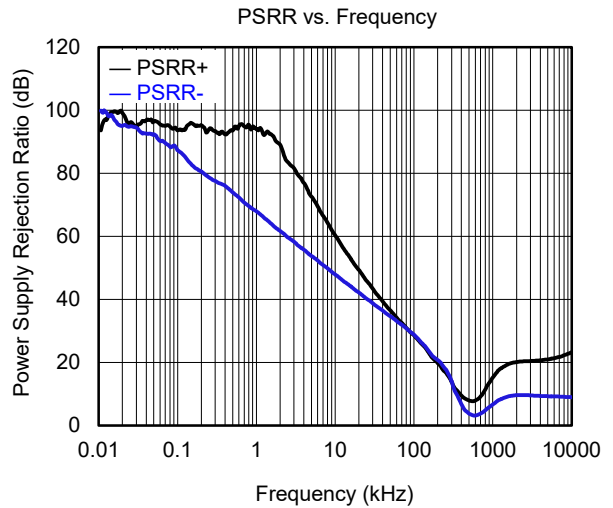
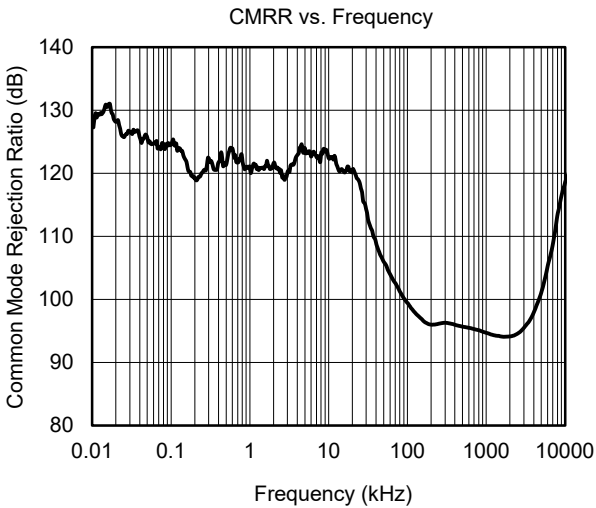
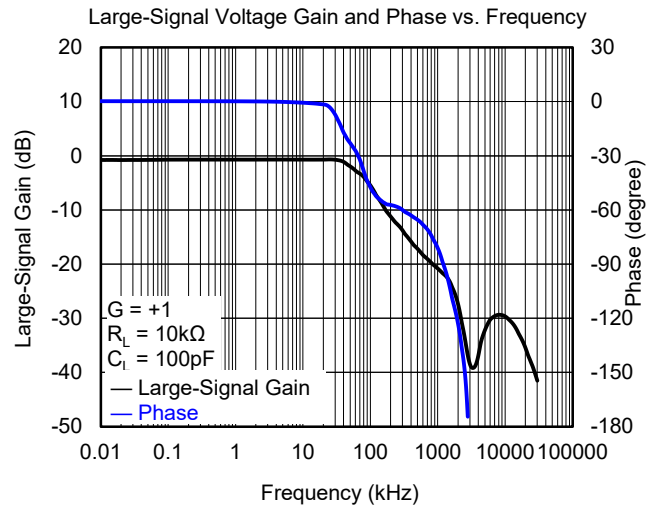
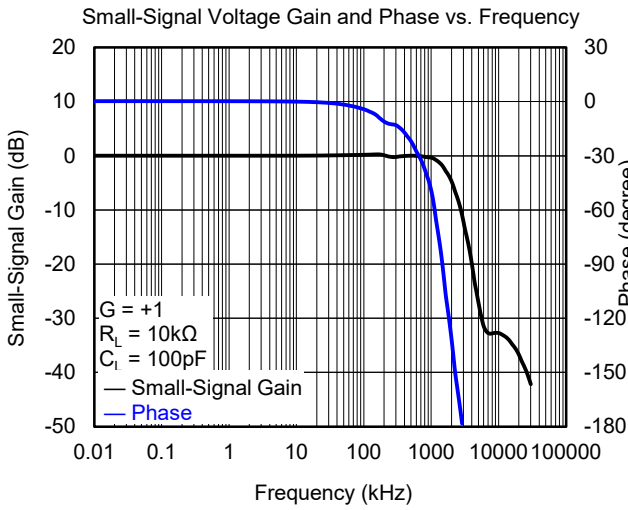
At  $T_A = +25^\circ\text{C}$ ,  $V_S = 3.3\text{V}$ ,  $V_{CM} = V_S/2$ ,  $V_{EN} = +V_S$ , unless otherwise noted.



# SGM8607-2 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At  $T_A = +25^\circ\text{C}$ ,  $V_S = 3.3\text{V}$ , unless otherwise noted.





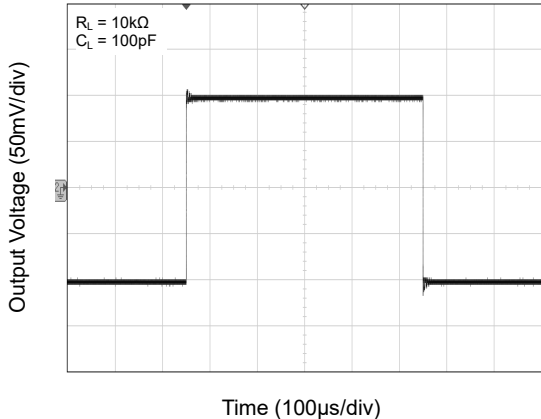
# 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## SGM8607-2

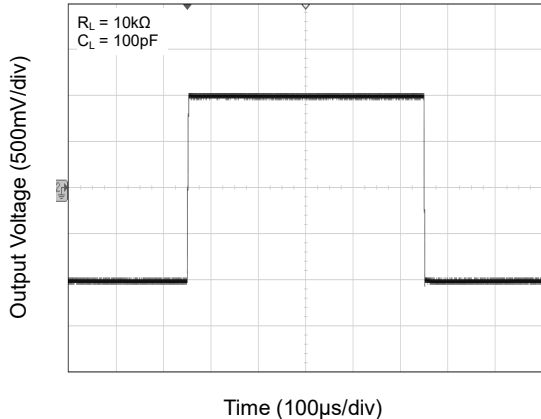
### TYPICAL PERFORMANCE CHARACTERISTICS (continued)

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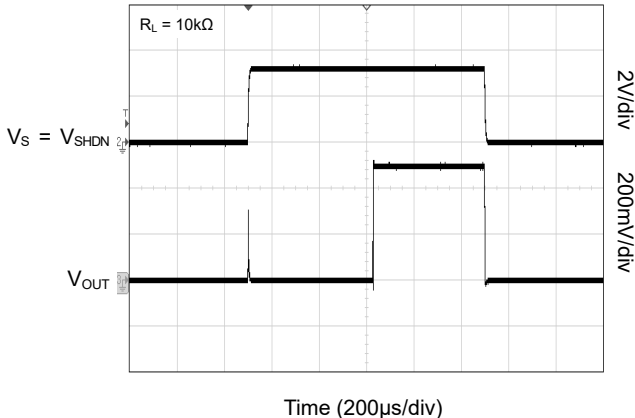
Small-Signal Step Response



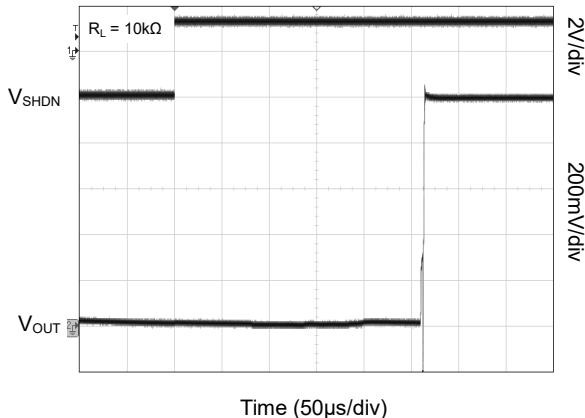
Large-Signal Step Response



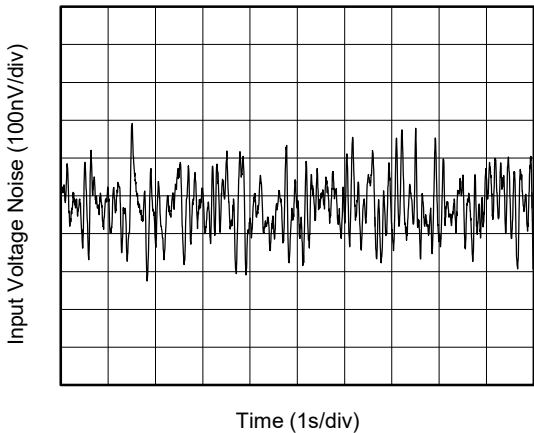
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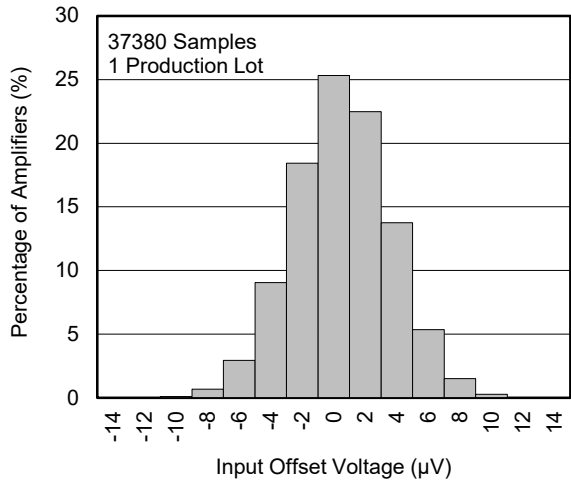
Shutdown Enable Response



0.1Hz to 10Hz Input Voltage Noise

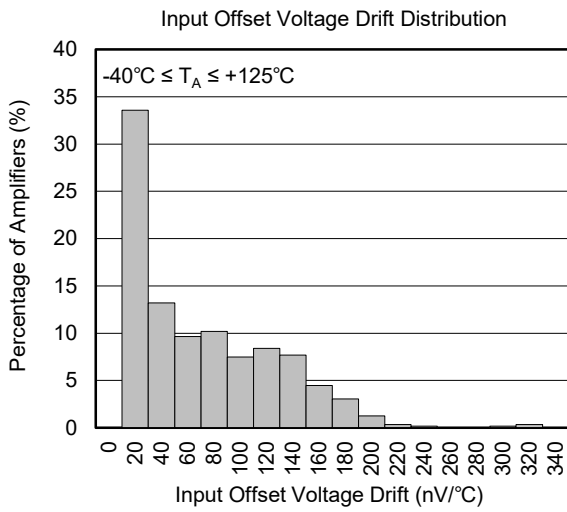


Input Offset Voltage Production Distribution



**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

At  $T_A = +25^\circ\text{C}$ ,  $V_S = 3.3\text{V}$ , unless otherwise noted.



# 1.2MHz, High Precision, Rail-to-Rail I/O, Dual CMOS Operational Amplifier in WLCSP Package

## SGM8607-2

### APPLICATION INFORMATION

#### Rail-to-Rail Input

The input common mode voltage range of the SGM8607-2 extends 100mV beyond the supply rails for the full supply voltage range. In Figure 1, the ESD diodes between the inputs and the power supply rails will clamp the input voltage not to exceed the rails.

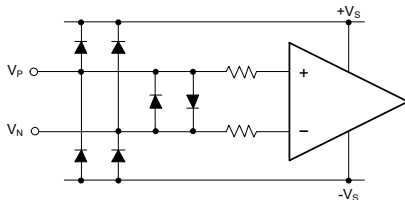


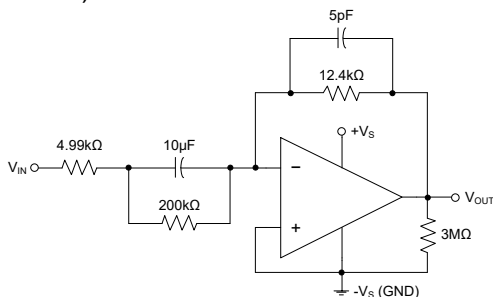
Figure 1. Input Equivalent Circuit

#### Input Current-Limit Protection

For ESD diode clamping protection, when the current flowing through ESD diode exceeds the maximum rating value, the ESD diode and amplifier will be damaged, so current-limit protection will be added in some applications. One resistor is selected to limit the current not to exceed the maximum rating value. In Figure 2, a series input resistor is used to limit the input current, but the drawback of this current-limit resistor is that it contributes thermal noise at the amplifier input. If this resistor must be added, its value must be selected as small as possible.

In addition, a clamping circuit on its inputs can sink current (up to 20mA) when the input is  $(+V_S) + 0.3V$ . This is for protecting from a worst  $V_{IN}$  short-circuit condition, and any alternate would need to have some sort of clamping behavior on its inputs. But in the shutdown mode, the clamping circuit does not work.

In the circuit (Figure 2), the operational amplifier inputs can handle a  $V_{IN}$  pulse (20V pulses width 1μs, rise and fall times 1ns).



NOTE: A current-limit resistor is required if the input voltage exceeds the supply rails by  $\geq 0.3V$ .

Figure 2. Input Current-Limit Protection

#### Rail-to-Rail Output

The SGM8607-2 supports rail-to-rail output operation. In single power supply application, for example, when  $+V_S = 3.3V$ ,  $-V_S = GND$ , 600Ω load resistor is tied from OUT pin to ground, the typical output swing range is from 0.032V to 3.25V.

#### Driving Capacitive Loads

The SGM8607-2 can directly drive a 1nF capacitor at Gain = +20 or a 100pF capacitor at Gain = +1 without oscillation. If greater capacitive load must be driven in application, the circuit in Figure 3 can be used. In this circuit, the IR drop voltage generated by  $R_{ISO}$  is compensated by feedback loop.

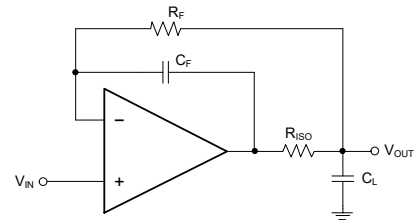


Figure 3. Circuit to Drive Heavy Capacitive Load

#### Power Supply Decoupling and Layout

A clean and low noise power supply is very important in amplifier circuit design. Besides of input signal noise, the power supply is one of important source of noise to the amplifiers through  $+V_S$  and  $-V_S$  pins. Power supply bypassing is an effective method to clear up the noise at power supply, and the low impedance path to ground of decoupling capacitor will bypass the noise to GND. In application, 10μF ceramic capacitor paralleled with 0.1μF or 0.01μF ceramic capacitor is used in Figure 4. The ceramic capacitors should be placed as close as possible to  $+V_S$  and  $-V_S$  power supply pins.

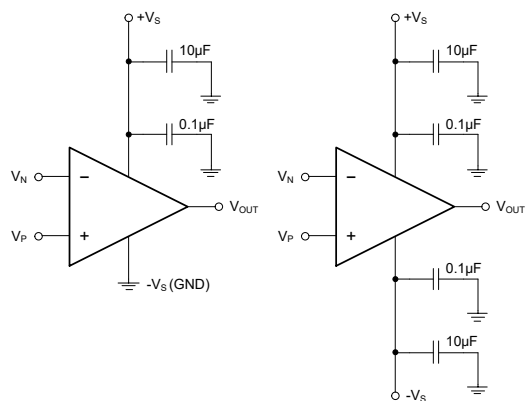


Figure 4. Amplifier Power Supply Bypassing

**APPLICATION INFORMATION (continued)**

**Grounding**

In low speed application, one node grounding technique is the simplest and most effective method to eliminate the noise generated by grounding. In high speed application, the general method to eliminate noise is to use a complete ground plane technique, and the whole ground plane will help distribute heat and reduce EMI noise pickup.

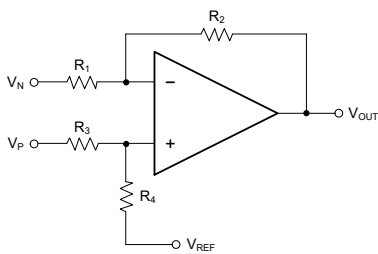
**Reduce Input-to-Output Coupling**

To reduce the input-to-output coupling, the input traces must be placed as far away from the power supply or output traces as possible. The sensitive trace must not be placed in parallel with the noisy trace in same layer. They must be placed perpendicularly in different layers to reduce the crosstalk. These PCB layout techniques will help to reduce unwanted positive feedback and noise.

**Typical Application Circuits**

**Difference Amplifier**

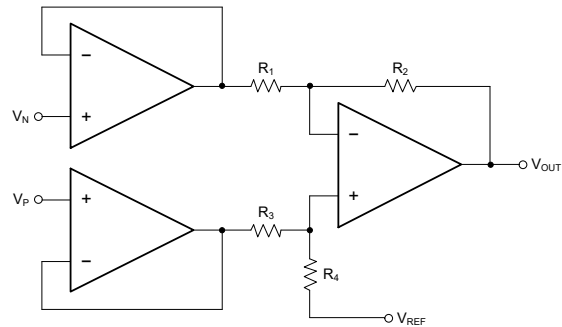
The circuit in Figure 5 is a design example of classical difference amplifier. If  $R_4/R_3 = R_2/R_1$ , then  $V_{OUT} = (V_P - V_N) \times R_2/R_1 + V_{REF}$ .



**Figure 5. Difference Amplifier**

**High Input Impedance Difference Amplifier**

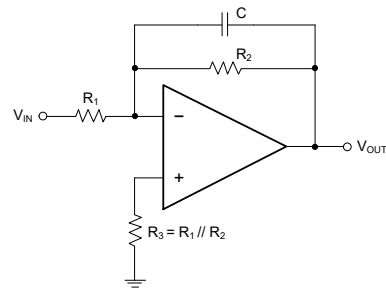
The circuit in Figure 6 is a design example of high input impedance difference amplifier. The added amplifiers at the input are used to increase the input impedance and eliminate drawback of low input impedance in Figure 5.



**Figure 6. High Input Impedance Difference Amplifier**

**Active Low-Pass Filter**

The circuit in Figure 7 is a design example of active low-pass filter, the DC gain is equal to  $-R_2/R_1$  and the -3dB corner frequency is equal to  $1/2\pi R_2 C$ . In this design, the filter bandwidth must be less than the bandwidth of the amplifier, and the resistor values must be selected as low as possible to reduce ringing or oscillation generated by the parasitic parameters in PCB layout.



**Figure 7. Active Low-Pass Filter**

**REVISION HISTORY**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

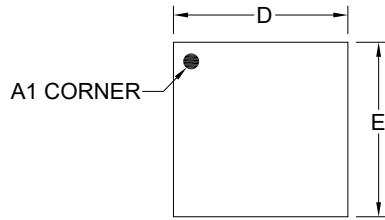
<b>Changes from Original (DECEMBER 2023) to REV.A</b>	<b>Page</b>
Changed from product preview to production data.....	All

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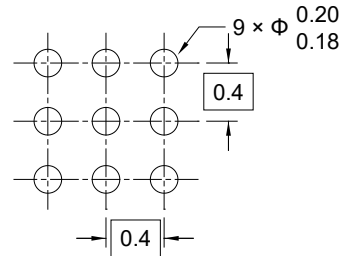
# PACKAGE INFORMATION

## PACKAGE OUTLINE DIMENSIONS

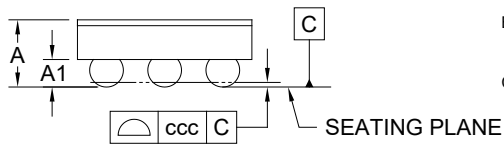
### WLCSP-1.2×1.2-9B-A



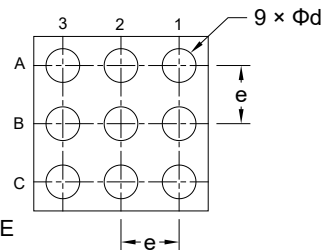
TOP VIEW



RECOMMENDED LAND PATTERN (Unit: mm)



SIDE VIEW



BOTTOM VIEW

Symbol	Dimensions In Millimeters		
	MIN	MOD	MAX
A	-	-	0.500
A1	0.169	-	0.209
D	1.170	-	1.230
E	1.170	-	1.230
d	0.202	-	0.262
e	0.400 BSC		
ccc	0.050		

NOTE: This drawing is subject to change without notice.

# PACKAGE INFORMATION

## TAPE AND REEL INFORMATION

### REEL DIMENSIONS



### TAPE DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

### KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
WLCSP-1.2×1.2-9B-A	7"	9.2	1.31	1.31	0.58	4.0	4.0	2.0	8.0	Q1

DD0001

# PACKAGE INFORMATION

## CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

## KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton
7" (Option)	368	227	224	8
7"	442	410	224	18

DD0002