

SGM8199N Voltage Output, High-side or Low-side Measurement, Bidirectional Current Shunt Monitor

GENERAL DESCRIPTION

The SGM8199N is a voltage output, bidirectional current shunt monitor, which is designed to sense voltage drops across the shunt at common mode voltage range without being affected by the supply voltage. The device features a wide common mode voltage range from -0.1V to 26V. It also offers a low offset, which allows 10mV full-scale maximum drops across the shunt when sensing current.

The SGM8199N provides a fixed gain of 100V/V. It operates from 2.7V to 26V single supply, consuming only 85μ A quiescent current.

The SGM8199N is available in a Green SC70-6 package. It is specified over the extended -40° C to $+125^{\circ}$ C temperature range.

FEATURES

- Wide Input Common Mode Range: -0.1V to 26V
- Low Input Offset Voltage: 350µV (V_{CM} = 12V, MAX) (Enable 10mV Full-Scale Shunt Drops)
- Gain: 100V/V
- Gain Error: ±0.4% (MAX)
- Low Quiescent Current: 85µA (TYP)
- -40°C to +125°C Operating Temperature Range
- Available in a Green SC70-6 Package

APPLICATIONS

Laptops Mobile Phones Telecom Equipment Power Managements Charging Systems Welding Equipment

TYPICAL APPLICATION





PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION	
SGM8199N	SC70-6	-40°C to +125°C	SGM8199NXC6G/TR	CEAXX	Tape and Reel, 3000	

MARKING INFORMATION

NOTE: XX = Date Code.

<u>YYY X X</u>

Date Code - Month

—— Date Code - Year —— Serial Number

Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	28V
V _{IN} Differential	±28V
Input Common Mode Voltage Range	GND - 0.3V to 28V
Input Current into All Pins	5mA
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 10s)	+260°C
ESD Susceptibility	
HBM	
CDM	2000V

RECOMMENDED OPERATING CONDITIONS

Operating Temperature Range-40°C to +125°C

OVERSTRESS CAUTION

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

ESD SENSITIVITY CAUTION

This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

PIN CONFIGURATION





ELECTRICAL CHARACTERISTICS

(At $T_A = +25^{\circ}C$, Full = -40°C to +125°C, $V_{CC} = 2.7V$ to 26V, $V_{IN+} = 12V$, $V_{SENSE} = V_{IN+} - V_{IN-}$, and $V_{REF} = V_{CC}/2$, unless otherwise noted.)

PARAMETER		CONDITIONS	TEMP	MIN	TYP	MAX	UNITS
Input Characteristics		•					
Input Offset Voltage, RTI ⁽¹⁾ (V _{os})		$\frac{V_{CM} = -0.1V \text{ to } 2V, V_{SENSE} = 0\text{mV}}{V_{CM} = 2V \text{ to } 26V, V_{SENSE} = 0\text{mV}}$			100	450	
					100	350	μν
Input Offset Voltage Drift (ΔV _{os} /Δ	T)		Full		1.5		µV/°C
Input Bias Current (I _B)		V _{SENSE} = 0mV	+25°C		16		μA
Input Offset Current (Ios)		V _{SENSE} = 0mV	+25°C		100		nA
Input Common Mode Voltage Rar	nge (V _{см})		Full	-0.1		26	V
		V_{CM} = -0.1V to 26V, V_{SENSE} = 0mV		89	108		<u> </u>
Common Mode Rejection Ratio (C	JMRR)	V_{CM} = 2V to 26V, V_{SENSE} = 0mV	Full	90	106		dB
Output Characteristics		•				1	4
Gain (G)			+25°C		100		V/V
Gain Error		V _{SENSE} = -5mV to 5mV	Full		0.05	0.4	%
Gain Temperature Coefficient			Full		15		ppm/°C
Nonlinearity Error		V _{SENSE} = -5mV to 5mV	+25°C		0.01		%
Maximum Capacitive Load		No sustained oscillation	+25°C	+25°C			nF
	V _{OH}				140	300	mV
Output voltage Swing from Rall	V _{OL}	$V_{CC} = 26V, R_L = 10K\Omega$ to $V_{CC}/2$	+25°C		90	200	
Frequency Response	•	•					<u>.</u>
-3dB Bandwidth		$V_{CC} = 5V, R_L = 10k\Omega, C_L = 10pF$	+25°C		20		kHz
		f = 1kHz, R _L = 10kΩ, V _{OUT} = 1V _{P-P}			0.04		N/har
Slew Rate (SR)		f = 1kHz, R_L = 10k Ω , V_{OUT} = 2 V_{P-P}	+25°C		0.09		η v/μs
Noise, RTI							<u>.</u>
Voltage Noise Density (e _n)		f = 1kHz	+25°C		50		nV/ _{√Hz}
Power Supply		•					<u>.</u>
Operating Voltage Range (V _{cc})			Full	2.7		26	V
Orden and Orment (L.)			+25°C		85	120	
Quiescent Current (IQ)		$v_{CC} = 26V, V_{SENSE} = UMV$	Full			130	μΑ
Power Supply Rejection Ratio (PSRR)		V_{CC} = 2.7V to 26V, V_{SENSE} = 0mV	+25°C		110		dB

NOTE: 1. RTI = Referred-to-input.

TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL PERFORMANCE CHARACTERISTICS (continued)



TYPICAL PERFORMANCE CHARACTERISTICS (continued)



TYPICAL PERFORMANCE CHARACTERISTICS (continued)





APPLICATION INFORMATION

Recommended Connection

The recommended connection of SGM8199N is shown in Figure 1. Also, it is recommended that the shunt resistor should be placed as close as possible to the two input pins of the current shunt monitor, so the extra resistance which is series with the measured shunt resistor will be reduced.

The bypass capacitor is necessary for the current shunt monitor to improve its stability. In most applications, the power supply is noisy and it will affect the operation of SGM8199N, so the bypass capacitor is needed in this case.

Power Supply

The SGM8199N can accurately measure the current when the common mode voltage exceeds the power supply voltage presented at the V_{CC} pin. For example, the V_{CC} power supply can be 5V and the load or common mode power supply voltage is allowed to reach up to 26V. The output voltage range is limited by the level of the power supply.

Selection of R_{SHUNT}

For the typical range of the differential input, the current shunt monitor of SGM8199N can work accurately with the order of 10mV. The application of SGM8199N will determine the selection of the shunt resistor R_{SHUNT} . Also, the users should consider the trade-off between voltage loss and the accuracy of small input signals. The effect of offset can be minimized by using high values of R_{SHUNT} , while the voltage loss can be minimized by using low values of R_{SHUNT} . For most applications, a voltage drop of 30mV over R_{SHUNT} is the appropriate range for the selection of R_{SHUNT} , and the corresponding offset voltage is 450µV only.

Unidirectional Application

In unidirectional operation, the current measuring direction of SGM8199N is fixed. Usually, the REF pin is connected directly to GND pin to ensure that the output is biased at 0V. Also, if the users desire to measure low input voltage with high accuracy, biasing REF pin to 200mV is a good method to set the current shunt monitor to its linear region.

For the least frequent case, REF pin can connect to the power supply pin directly to measure the negative current. In this case, the output voltage is equal to the voltage of power supply when input voltage is equal to 0mV.

Bidirectional Application

Bidirectional application indicates that the current shunt monitor of SGM8199N can measure the current for two different directions. For this special case, the REF pin can be set anywhere between 0V to V_{CC} for the bias of output voltage. For simplification, it is recommended that $V_{REF} = V_{CC}/2$ is a typical voltage point. However, if the absolute values of positive and negative current are not equal, the voltage of REF pin should be set other than $V_{CC}/2$.

Input Filtering

It is not recommended to add a filter at the output of SGM8199N, as doing so increases impedance seen at the output of the internal buffer. Filtering at the input pins will be a good choice as long as the change of the input impedance is taken in account. Figure 2 indicates the application of input filtering.

For minimizing the error and enhancing the accuracy of the result, the resistance of the external resistor R_S should be less than 10Ω . At the internal input structure of SGM8199N, there is a bias network which causes mismatch in the bias current I_B of two input pins, and the external resistors result in the mismatch of voltage drops (because of the mismatch of I_B) across R_S so that the differential error will be created. Also, this differential error will be reflected to the input pins of the current shunt monitor and result in the accuracy. However, the bias current difference can result in little effect on the operation of the current shunt monitor so the users do not need to care about that.



Figure 2. Filtering at Input

APPLICATION INFORMATION (continued)

Shutting Down SGM8199N

There is no shutdown control at the internal of SGM8199N, so the only way to shut down the quiescent current from the power supply is by using an external logic gate or transistor switch.

However, the users may concern about the amount of current through SGM8199N in shutdown mode. The schematic in Figure 3 can be used to evaluate the amount of current drain in shutdown mode.



NOTE: $1M\Omega$ path from shunt inputs to reference and SGM8199N output.

Figure 3. Shutting Down the Current Shunt Monitor with V_{REF} = 0

Input Impedance of REF Input

The common mode rejection ratio (CMRR) can be influenced by the input impedance of REF Pin, but the users do not need to concern about it if REF pin is driven by a power supply. However, if the REF pin is driven by resistive dividers, it should be buffered by an operational amplifier to provide low input impedance.

If the output can be measured differentially, for example, a differential analog-to-digital converter, the users do not need to care about the external impedance which is added at the REF pin and the effect can be removed. Figure 4 indicates an example about how to remove the effect of extra impedance at REF pin.



Figure 4. Cancelling the Input Impedance at REF Input



APPLICATION INFORMATION (continued)

The Performance of SGM8199N when Common Mode Transients above 26V

The SGM8199N can be powered by the power supply whose power-on transient is higher than 26V, especially for the application of automotive industry. At this case, a Zener diode or Zener-type transient absorber (Transzorbs) can be used to prevent the current shunt monitor from over-voltage during the transient of power-on. It does not recommend the users to use the transient absorber other than Transzorbs because of the significant time delay. However, any Zener diode needs extra operational resistors to provide the operating current, so the 10 Ω resistors are selected (any large external resistor will affect the gain). Also, 10 Ω resistors and a Zener diode with lowest rating of power are enough for handling the short-term transient in most applications.



Figure 5. Using Dual Diodes for Absorbing Supply Transient

If the low power Zener diode cannot protect the current shunt monitor from supply transient, a Transzorb with high power must be taken into account in this case. In Figure 6, for saving the space of PCB board, a Transzorb and back-to-back diode can also be used to absorb the transient. For the applications of Figure 5 and Figure 6, the total area which is occupied by SGM8199N itself and its protective components is equivalent to the MSOP-8 package, and smaller than SOIC-8 package.



Figure 6. Using Transzorb and Input Clamps for Absorbing Supply Transient

REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

OCTOBER 2020 – REV.A to REV.A.1	Page
Updated Typical Performance Characteristics section	
Changes from Original (AUGUST 2019) to REV.A	Page
Changed from product preview to production data	All



PACKAGE OUTLINE DIMENSIONS

SC70-6





RECOMMENDED LAND PATTERN (Unit: mm)





Symbol	Dimer In Milli	nsions meters	Dimensions In Inches		
	MIN	MAX	MIN	MAX	
A	0.800	1.100	0.031	0.043	
A1	0.000	0.100	0.000	0.004	
A2	0.800	1.000	0.031	0.039	
b	0.150	0.350	0.006	0.014	
с	0.080	0.220	0.003	0.009	
D	2.000	2.200	0.079	0.087	
E	1.150	1.350	0.045	0.053	
E1	2.150	2.450	0.085	0.096	
е	0.65	TYP	0.026 TYP		
e1	1.300) BSC	0.051 BSC		
L	0.525 REF		0.021	REF	
L1	0.260	0.460	0.010	0.018	
θ	0°	8°	0°		

TAPE AND REEL INFORMATION

REEL DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SC70-6	7″	9.5	2.40	2.50	1.20	4.0	4.0	2.0	8.0	Q3

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton	
7" (Option)	368	227	224	8	
7"	442	410	224	18	

